



AIASYB2:



APLICACIÓN DE LA INTELIGENCIA ARTIFICIAL EN LOS
SENSORES Y BIOSENSORES

Multiresolution Analysis Wavelets

Jesús Manuel de la Cruz García, Matilde Santos Peña
Dpto. Arquitectura de Computadores y Automática.
Universidad Complutense

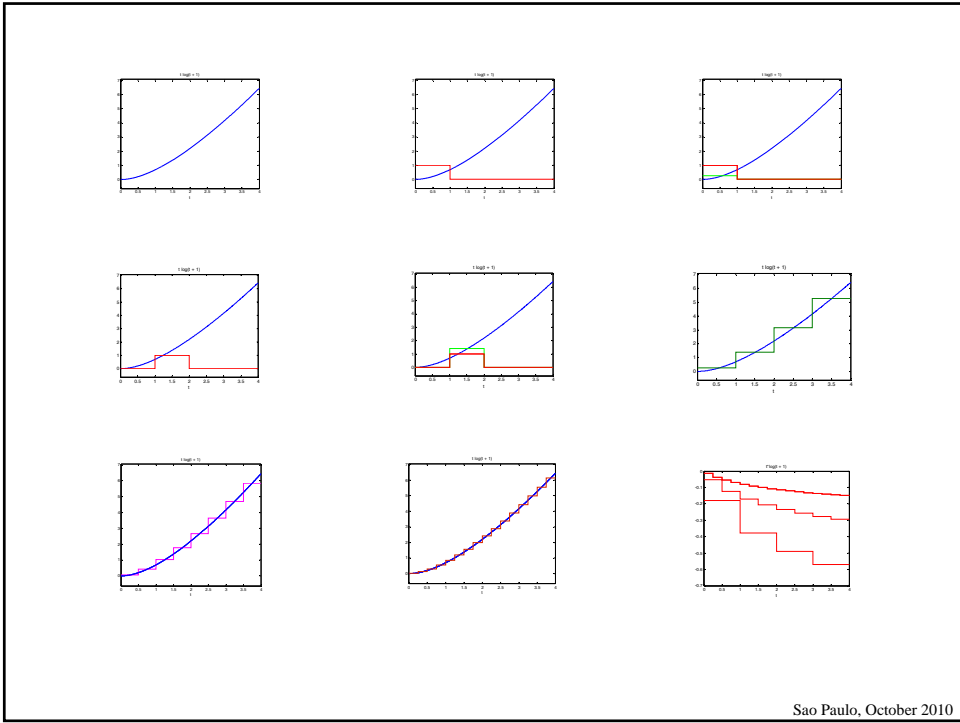
jmcruz@fis.ucm.es, msantos@dacya.ucm.es

<http://www.dacya.ucm.es/area-isa/index.php?page=home>

Sao Paulo, Brasil, October 2010

Multiresolution Analysis

- Example of Scaling and Wavelet Functions
- Nested Spaces and Complementary Spaces
- Multiresolution
- Fourier Transform versus Wavelet Transform
- Discrete Wavelets Transform
- Applications
- Bibliography

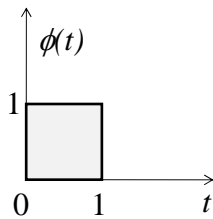
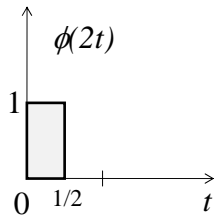


Scaling Functions

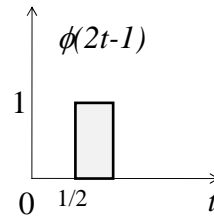
- Continuous time box function: $\phi(t)$

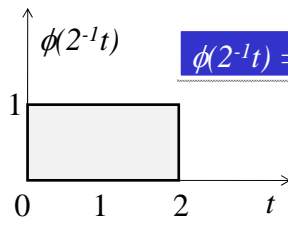
$$\phi(t) = \phi(2t) + \phi(2t-1)$$

Scaling (compressing)



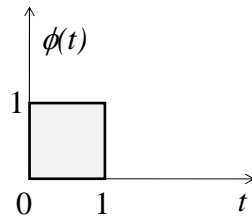
Scaling + Shifting



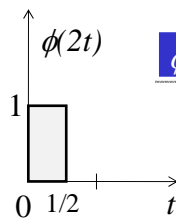
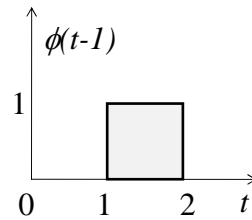


$$\phi(2^{-1}t) = \phi(t) + \phi(t-1)$$

Scaling

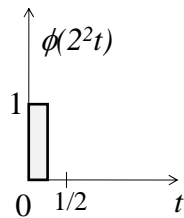


Scaling + Shifting

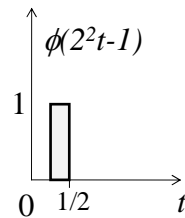


$$\phi(2t) = \phi(2^2t) + \phi(2^2t-1)$$

Scaling



Scaling + Shifting



For this function

$$\phi(2^{-1}t) = \phi(t) + \phi(t-1)$$

$$\phi(t) = \phi(2t) + \phi(2t-1)$$

$$\phi(2t) = \phi(2^2t) + \phi(2^2t-1)$$

We can generalize :

$$\phi(2^{-1}t) = 2 \sum_{k=0}^N h(k) \phi(t-k)$$

$$\phi(t) = 2 \sum_{k=0}^N h(k) \phi(2t-k) \quad \text{For the box function } h(0)=h(1)=1/2$$

$$\phi(2t) = 2 \sum_{k=0}^N h(k) \phi(2^2t-k)$$

$\phi(t)$ is called a **scaling function**

$$\phi(t) = 2 \sum_{k=0}^N h(k) \phi(2t-k) \quad \text{Refinement equation}$$

- This equation couples the representations of a continuous time function at two time scales.
- The continuous time function is determined by a discrete time filter

$$H(z) = h(0) + h(1)z^{-1} + \dots + h(N)z^{-N}$$

- For the example :

$$h(0) = h(1) = 1/2 \quad (\text{lowpass filter})$$

- **Solutions to the refinement equation may not always exist.**

If it does.

- $\phi(t)$ has a compact support i.e.
 $\phi(t) = 0$ if $t < 0$ or $t \geq N$ (~ it has compact support)
 (comes from the FIR filter, $h(N)$)

- $\phi(t)$ often has no closed form solution
- $\phi(t)$ is unlikely to be smooth

- **Constraint on $h(k)$:** $\int \phi(t) dt = 2 \sum_{k=0}^N h(k) \int \phi(2t-k) dt =$
 $= 2 \sum_{k=0}^N h(k) \frac{1}{2} \int \phi(t) dt$

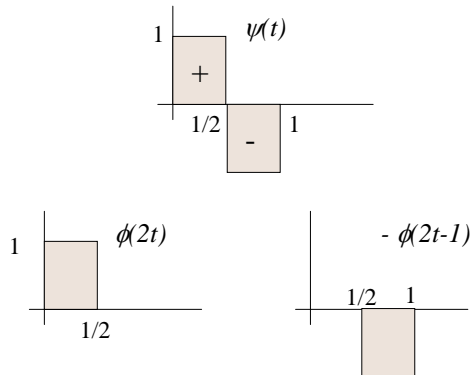
then

$$\sum_{k=0}^N h(k) = 1$$

(assuming $\int \phi(t) dt \neq 0$)

$$\phi(t) = 2 \sum_{k=0}^N h(k) \phi(2t-k)$$

Wavelet de Haar



$$\psi(t) = \phi(2t) - \phi(2t-1)$$

- More generally:

$$\psi(t) = 2 \sum_{k=0}^N g(k) \phi(2t - k) \quad \text{Wavelet equation}$$

- The wavelet is determined by a discrete time filter

$$G(z) = g(0) + g(1) z^{-1} + \dots + g(N) z^{-N}$$

For the Haar wavelet example:

$$g(0) = 1/2, \quad g(1) = -1/2$$

Frequency response for filters $H(z)$ y $G(z)$

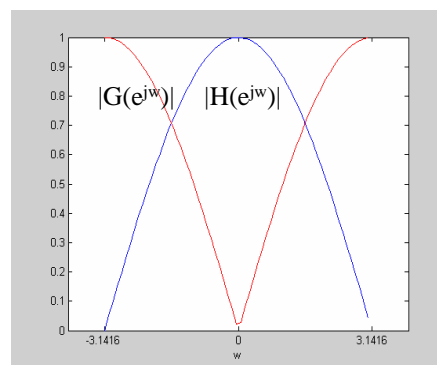
$H(z)$: lowpass filter

$G(z)$: highpass filter

Haar:

$$H(z) = 1/2 + 1/2 z^{-1}$$

$$G(z) = 1/2 - 1/2 z^{-1}$$



Orthogonality of Scaling Functions

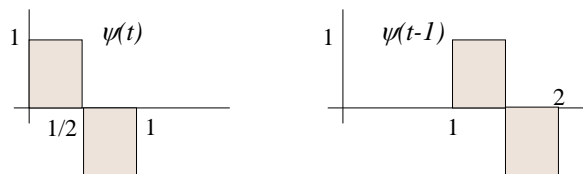
1. Orthogonality under integer shifts



$$\int \phi(t)\phi(t-k) dt = \delta(k) = \begin{cases} 1 & \text{if } k=0 \\ 0 & \text{otherwise} \end{cases}$$

Orthogonality of Wavelet Functions

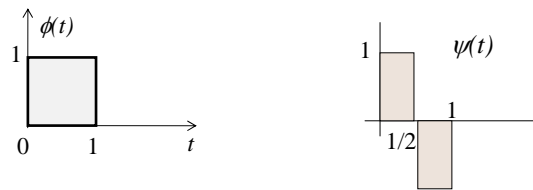
1. Wavelet are orthogonal under integer shifts



$$\int \psi(t)\psi(t-k) dt = \delta(k)$$

Orthogonality of Scaling and Wavelet

2. Scaling function is orthogonal to wavelet

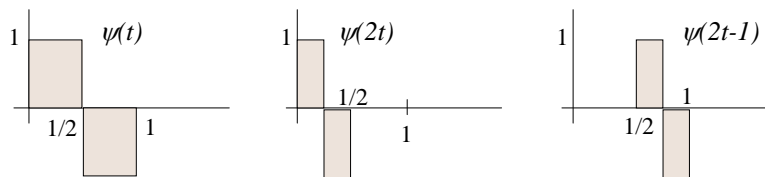


$$\int \phi(t)\psi(t) dt = 0$$

Positive and negative areas cancel each other

Orthogonality of Wavelet across Scales

3. Wavelet are orthogonal across scales



$$\int \psi(t)\psi(2t) dt = 0, \quad \int \psi(t)\psi(2t-1) dt = 0$$

Bases

- Our goal is to use the wavelet function $\psi(t)$, its scaled versions, $\psi(st)$, and their shifts, $\psi(st-k)$, as building blocks for continuous time functions.
- We move with functions where a measurement has been introduced through a scale product:

$$f(t) \cdot g(t) \equiv \langle f(t), g(t) \rangle \equiv \int_{-\infty}^{\infty} f(t) g(t) dt < \infty$$

- Such functions must have finite energy, and they are said to belong to the Hilbert space, $L^2(\mathbb{R})$.

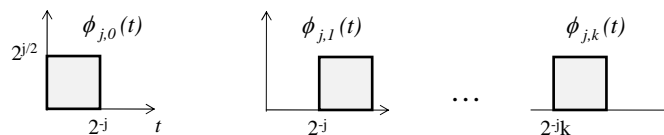
$$f(t) \cdot f(t) \equiv \|f(t)\|^2 = \int_{-\infty}^{\infty} f(t)^2 dt < \infty$$

Families of Orthonormal Scaling Functions

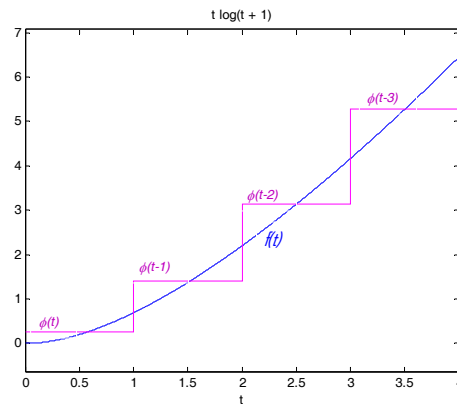
For every given scale $s^{-j} = 2^j$, j integer, we have a family of orthonormal functions

$$\{\phi_{j,k}(t)\}_{k: -\infty}^{\infty} = \{2^{j/2} \phi(2^j t - k)\}_{k: -\infty}^{\infty}, \quad -\infty \leq j \leq \infty$$

Normalization factor so that $\|\phi_{j,k}(t)\| = 1$

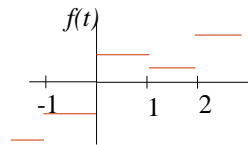


Relationship between $f(t) \in L^2(\mathbb{R})$ and $\phi_{0,k}(t)$



Every $\phi_{0,k}(t)$ contributes with the mean value of $f(t)$ in the interval $(k-1, k)$

The family $\{\phi(t-k)\}_{k \in \mathbb{Z}}$ generates a subspace V_0 of step functions in the intervals $t \in [k, k+1)$:

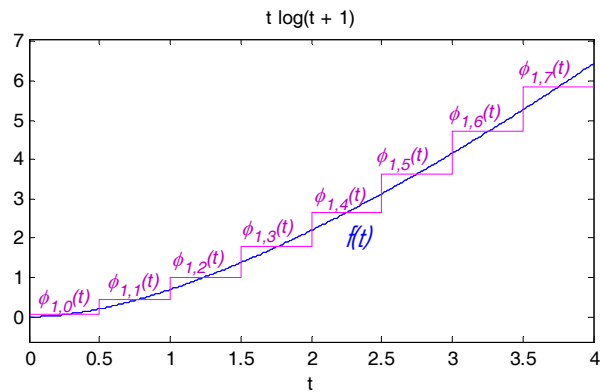


If $f(t) \in L^2(\mathbb{R})$ and $f(t) \in V_0$ then

$$f(t) = \sum_{k=-\infty}^{\infty} \langle f, \phi_{0,k} \rangle \phi_{0,k}$$

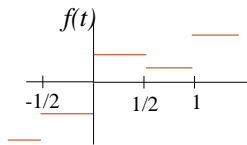
$\{\phi(t-k)\}_{k \in \mathbb{Z}}$ is an orthonormal bases of V_0

Relationship between $f(t) \in L^2(\mathbb{R})$ and $\phi_{1,k}(t)$



As the scale factor increases, the details increase

The family $\{\phi_{1,k}(t)\}_{k \in \mathbb{Z}}$ generates a subspace V_1 of step functions in the intervals $t \in 2^{-1}[k, k+1)$:

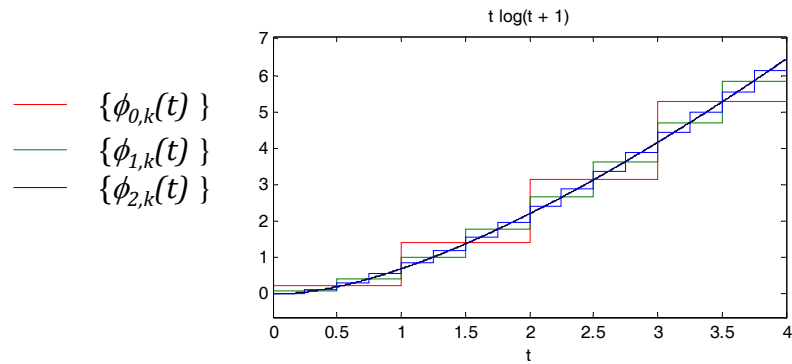


If $f(t) \in L^2(\mathbb{R})$ and $f(t) \in V_1$ then

$$f(t) = \sum_{k=-\infty}^{\infty} \langle f, \phi_{1,k} \rangle \phi_{1,k}$$

$\{\phi_{1,k}(t)\}_{k \in \mathbb{Z}}$ is an orthonormal bases of $V_1 \supset V_0$

Relationship between $f(t) \in L^2(\mathbb{R})$ and $\phi_{j,k}(t)$



Subspace V_j of step functions in the intervals $t \in 2^{-j}[k, k+1)$:

$\{\phi_{j,k}(t)\}_{k \in \mathbb{Z}}$ is an orthonormal bases of $V_j \supset V_{j-1} \dots \supset V_0$

But

$\{\phi_{j,k}(t)\}_{j, k \in \mathbb{Z}}$ is not an orthonormal basis of $L^2(\mathbb{R})$

However

$\{\psi_{j,k}(t)\}_{j, k \in \mathbb{Z}}$ is an orthonormal basis of $L^2(\mathbb{R})$

Family of Orthonormal Wavelet Functions

Consider all scales $s^{-l} = 2^j$, j integer, and integer shifts k of the Haar wavelet

$$\psi_{j,k}(t) = 2^{j/2} \psi(2^j t - k), \quad -\infty \leq j \leq \infty, \quad -\infty \leq k \leq \infty$$

Normalization factor so that $\|\psi_{j,k}(t)\| = 1$

$$\int \psi_{j,k}(t) \psi_{l,m}(t) dt = \begin{cases} 1 & \text{if } j = l \text{ and } k = m \\ 0 & \text{otherwise} \end{cases} \\ = \delta(j-l) \delta(k-m)$$

$\{\psi_{j,k}(t)\} \ j, k \in \mathbb{Z}$ is an orthonormal basis of $L^2(\mathbb{R})$

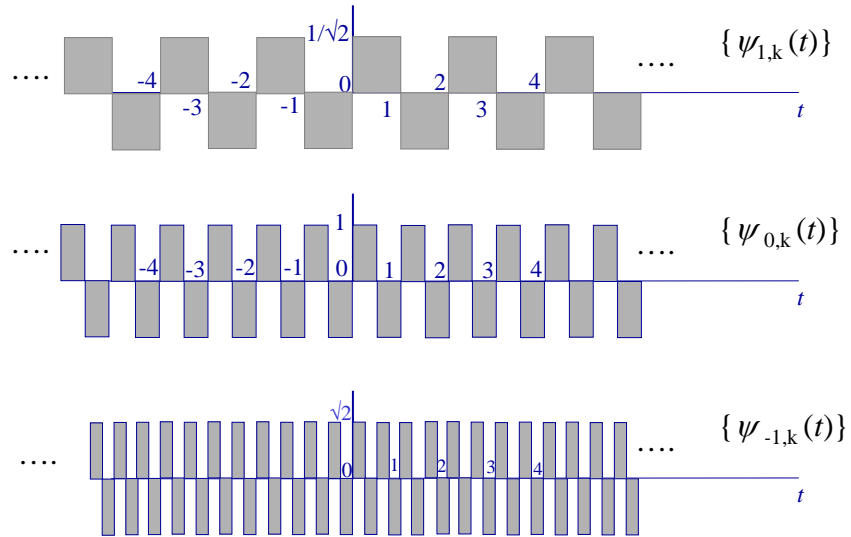
$$f(t) = \sum_{j,k \in \mathbb{Z}} b_{j,k} \psi_{j,k}(t)$$

$$b_{j,k} = \int_{-\infty}^{\infty} f(t) \psi_{j,k}(t) dt$$

$$\psi_{j,k}(t) = 2^{j/2} \psi(2^j t - k)$$

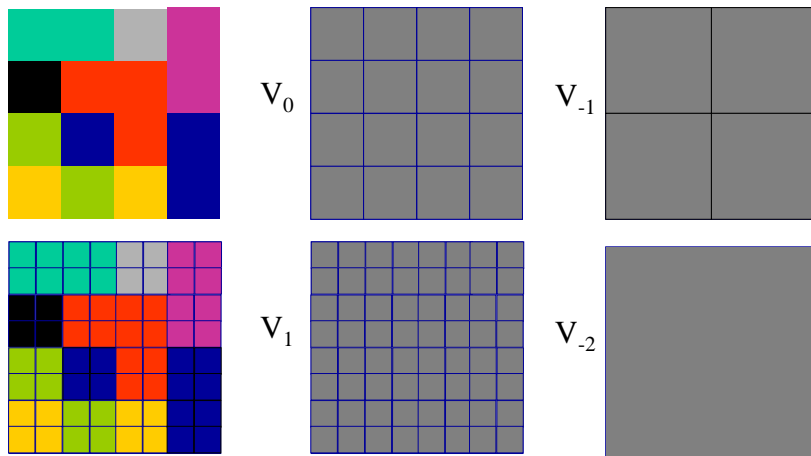
We need a two times infinity number of coefficients

Family of Orthonormal Wavelet Functions



Example of Similarity with Images

Every cell is an element of the basis



Multiresolution Analysis

In order to make a multiresolution analysis we need a sequence of embedded subspaces that verify:

1. Inclusion:

$$\mathbf{L}^2(\mathbf{R}) \supset \dots \supset V_{J+1} \supset V_J \supset \dots V_1 \supset V_0 \supset V_{-1} \supset \dots \supset \{0\}$$

2. Completeness: $\lim_{J \rightarrow \infty} V_J = \bigcup_{J=-\infty}^{\infty} V_J = \mathbf{L}^2(\mathbf{R})$

3. Emptiness: $\bigcap_{J=-\infty}^{\infty} V_J = \{0\}$

29

4. Shift: If $f(t) \in V_J$ then $f(t - 2^j k) \in V_J$

5. Scale: If $f(t) \in V_J$ entonces $f(2t) \in V_{J+1}$

6. Basis of V_0 : There exists $\phi(t)$ such that $\{\phi(t-k)\}_{k \in \mathbf{Z}}$ is a basis of V_0

A sequence of closed subspaces $\{V_J\}_{J \in \mathbf{Z}}$ verifying

1- 6 are called a **multiresolution approximation of $\mathbf{L}^2(\mathbf{R})$**

30

$V_{J+1} \supset V_J$: How, do we fill the gap between them?

Define a sequence of complementary subspaces W_j such that:

$$V_{J+1} = V_J + W_J$$

and they do not overlap

$$V_J \cap W_J = \{0\}$$

That is verified by Haar $\{\psi_{j,k}(t)\}_{k \in \mathbb{Z}}$ and $\{\phi_{j,k}(t)\}_{k \in \mathbb{Z}}$

Define W_j as the subspace generated by the orthonormal set

$$W_j : \left\{ \psi_{j,k}(t) = 2^{j/2} \psi(2^j t - k) \right\}_{k \in \mathbb{Z}}$$

Where:

$$\psi_{j,k}(t) = 2^{j/2} \psi(2^j t - k), \quad -\infty \leq j \leq \infty, \quad -\infty \leq k \leq \infty$$

is an orthonormal basis of $L^2(\mathbb{R})$. Then

$$\{\psi_{j \leq J, k}(t)\}_{(j,k) \in \mathbb{Z}} = \{\psi_{j < J, k}(t)\}_{(j,k) \in \mathbb{Z}} \cup \{\psi_{J, k}(t)\}_{k \in \mathbb{Z}}$$

Therefore: $V_J \perp W_J$

and $V_{J+1} = V_J \oplus W_J$

We then have:

$$V_1 = V_0 \oplus W_0$$

$$V_2 = V_1 \oplus W_1 = V_0 \oplus W_0 \oplus W_1$$

....

$$\mathbf{L}^2(\mathbf{R}) = V_0 \oplus \sum_{j=0}^{\infty} W_j$$

Moreover

$$V_0 = V_{-1} \oplus W_{-1}$$

$$= V_{-2} \oplus W_{-2} \oplus W_{-1}$$

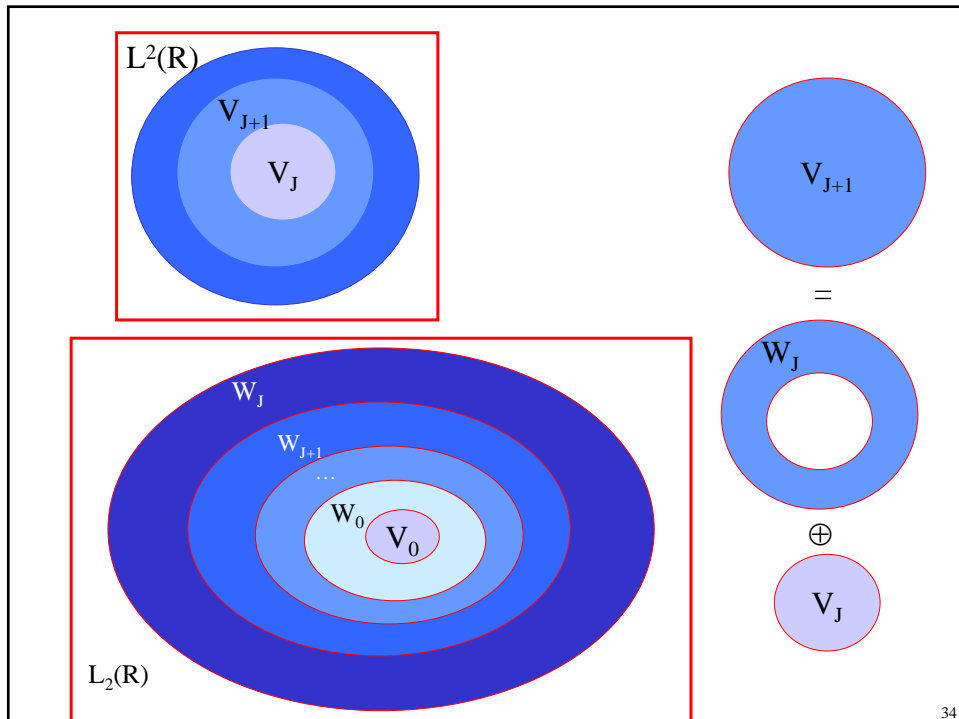
....

$$= \sum_{j=-\infty}^{-1} W_j$$

then:

$$\mathbf{L}^2(\mathbf{R}) = V_0 \oplus \sum_{j=0}^{\infty} W_j = \sum_{j=-\infty}^{\infty} W_j$$

33



34

Relationship between subspaces

We have

V_0 has a shift invariant basis $\{\phi(t-k)\}_{k \in \mathbb{Z}}$

W_0 has a shift invariant basis $\{\psi(t-k)\}_{k \in \mathbb{Z}}$

Since $V_1 = V_0 + W_0$ we expect that V_1 will have twice as many basis functions as V_0 alone.

Two possibilities:

1. $\{\phi(t-k), \psi(t-k)\}_{k \in \mathbb{Z}}$

2. Use the scaling law

If $\phi(t-k) \in V_0$ then $\phi(2t-k) \in V_1$

2. Use the scaling law: $\phi(t-k) \in V_0$, $\phi(2t-k) \in V_1$

So V_1 has a shift-invariant basis: $\{2^{-1/2}\phi(2t-k)\}_{k \in \mathbb{Z}}$

Since $V_0 \subset V_1$ any function in V_0 can be written as a linear combination of the basic functions for V_1

Then:

$$\phi(t) = 2 \sum_{k=0}^N h(k) \phi(2t-k) \quad \text{Refinement equation}$$

We also know:

$$W_0 = V_1 - V_0$$

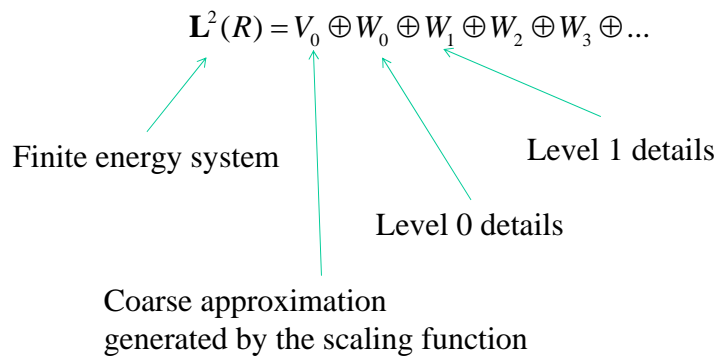
So $W_0 \subset V_1$ any function in W_0 can be written as a linear combination of the basic functions for V_1

Then:

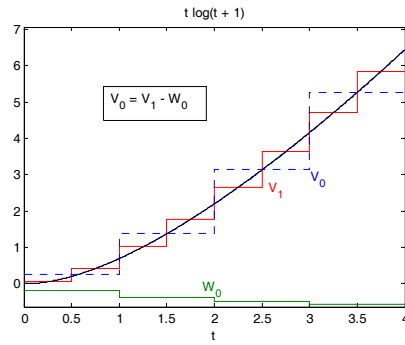
$$\psi(t) = 2 \sum_{k=0}^N g(k) \phi(2t - k)$$

Wavelet equation

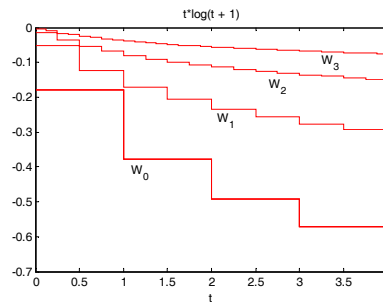
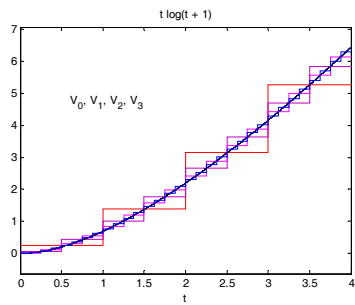
Multiresolution Representations



$$V_0 = V_J - W_{J-1} - W_{J-1} \cdots - W_0$$



$$V_0 = V_J - W_{J-1} - W_{J-1} \cdots - W_0$$

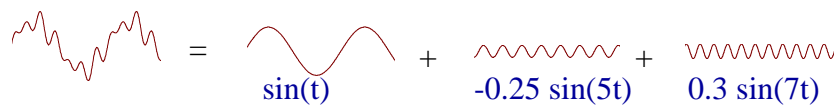


Fourier Transform versus Wavelet Transform

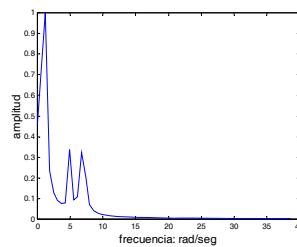
Sao Paulo, October 2010

Fourier Analysis

- Breaks down a signal into constituent sinusoids of different frequencies


$$\text{Signal} = \sin(t) + -0.25 \sin(5t) + 0.3 \sin(7t)$$

- Transform our view of the signal from from time-based to frequency-based.



- In transforming to the frequency domain, time information is lost:
 - When did a determined event took place?
- If it is a *stationary* signal this drawback isn't very important.
- Fourier analysis is not suited to detecting nonstationary or transitory characteristics:
 - drift,
 - trends,
 - abrupt changes: breakdown points, discontinuities in higher derivatives
 - beginnings and ends of events
 - self similarities.

Why Wavelets?

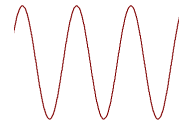
- Wavelet analysis allows the use of long time intervals where we want more precise low-frequency information, and shorter regions where we want high-frequency information (all in the same signal)
- Ability to perform *local analysis* : to analyze a localized area of a larger signal.
- Compress or de-noise a signal without appreciable degradation.

What is a Wavelet?

- A wavelet is a waveform of effectively limited duration that has an average value of zero.



- **Sinusoids** : unlimited duration, smooth and predictable.



- **Wavelets**: limited duration, irregular and asymmetric.

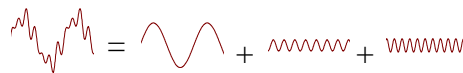


Continuos Wavelet Transform

- Fourier transform: breaks down a signal in sum of sinusoids of different frequencies → **Fourier Coefficients**

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$$



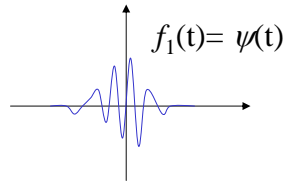
- Wavelet: breaks down a signal in sum of scaled and shifted versions of the wavelet function → **Wavelet Coefficients**

$$C(\text{scale}, \text{position}) = \int_{-\infty}^{\infty} f(t) \psi(\text{scale}, \text{position}) dt$$

$$f(t) = \iint C(\text{scale}, \text{position}) \psi^*(\text{scale}, \text{position}) de dp$$

C: measurement of similarity between the signal and the wavelet

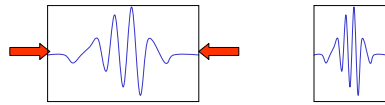
Scaling



$$s = 1/a$$

$$a = 2$$

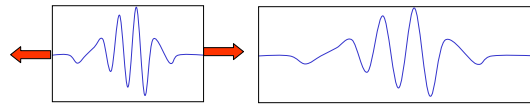
$$f_2(t) = \psi(2t)$$



A low scale compresses the signal \Rightarrow Fast changing \Rightarrow High frequencies

$$a = 1/2$$

$$f_3(t) = \psi(t/2)$$

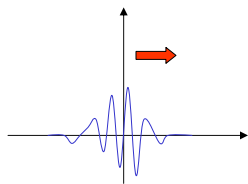


A high scale stretches the signal \Rightarrow Slow changing \Rightarrow Low frequencies

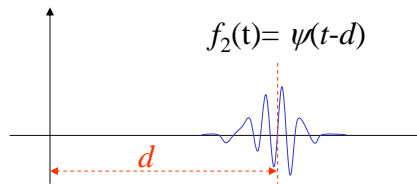
$$f_1(1) = f_2(0.5) = f_3(2)$$

Shifting

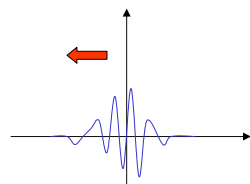
$$f_1(t) = \psi(t)$$



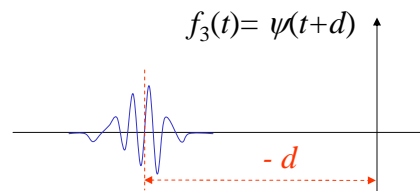
$$f_2(t) = \psi(t-d)$$



$$f_1(t) = \psi(t)$$



$$f_3(t) = \psi(t+d)$$



$$\text{Si } d = 5, f_1(0) = f_2(5) = f_3(-5)$$

Shifting

$$\psi(t-d)$$

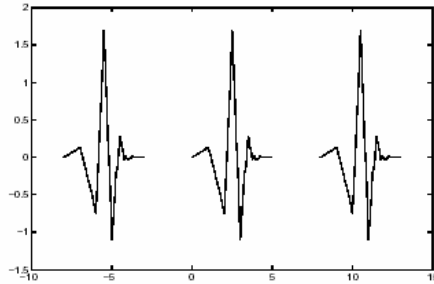
Scaling

$$\frac{1}{\sqrt{s}}\psi\left(\frac{t}{s}\right)$$

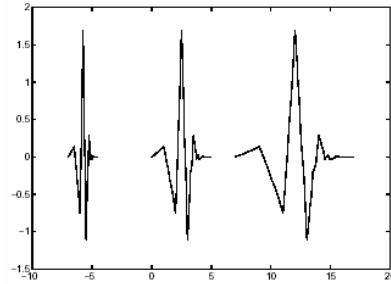
Scaling and Shifting

$$\frac{1}{\sqrt{s}}\psi\left(\frac{t-d}{s}\right)$$

$$C(s, d) = \int_{-\infty}^{\infty} f(t) \frac{1}{\sqrt{s}} \psi\left(\frac{t-d}{s}\right) dt$$



wavelet db3(t): centered
wavelet db3(t + 8): left
wavelet db3(t - 8): right



wavelet db3(t): centered
wavelet db3(2t+7): left
wavelet db3(t/2 - 7): right

Wavelet Properties

- Mother Wavelet:

$$\psi(t)$$

- Scaling and Shifting:

$$\psi_{s,d}(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t-d}{s}\right)$$

- Null mean value:

$$\int \psi(t) dt = 0$$

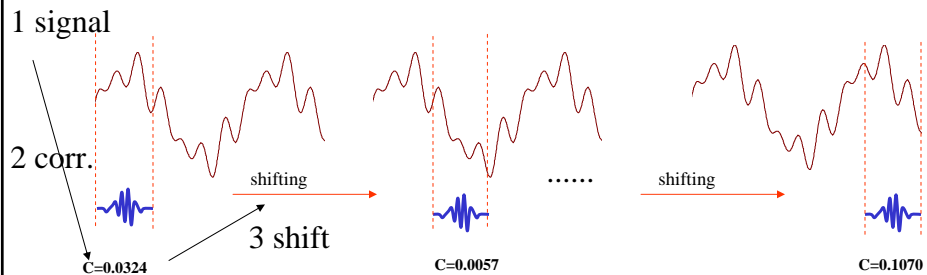
- Admissibility condition:
(wave of limited wide)

$$\int \frac{|\Psi(w)|^2}{|w|} dw < \infty, \Rightarrow |\Psi(0)|^2 = 0$$

- Regularity condition:
(concentrated in time)

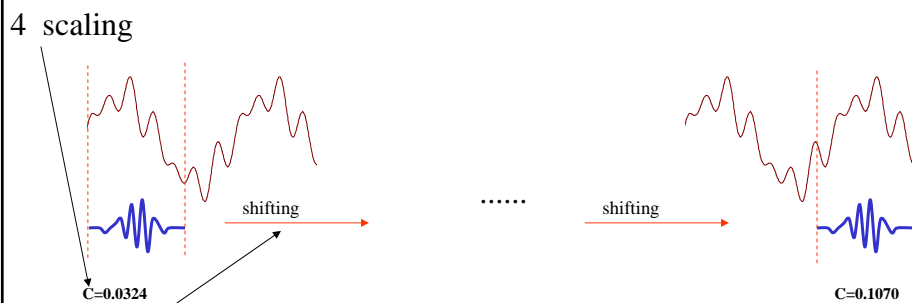
$$\int t^p \psi(t) dt \neq 0, \quad p = 0, 1, \dots, N$$

Steps to Compute the Coefficients



1. Take a wavelet and compare it to a section at the start of the original signal
2. Calculate a number, C , that represents how closely correlated the wavelet is with this section of the signal
3. Shift the wavelet to the right and repeat steps 1 and 2 until you've covered the whole signal.

Steps to Compute the Coefficients



4. Scale (stretch) the wavelet and repeat steps 1 through 3
5. Repeat steps 1 through 4 for all scales

Discrete Wavelet Transform

- Scale and displacement are continuous variables
- We choose only a finite subset of scales and displacement
- *Discrete wavelet transform:*
 - Displacements and scales in powers of 2:

$$s^{-1} = 2^j, \quad d = k 2^j = k s^{-1}, \quad j \text{ and } k \text{ integers}$$

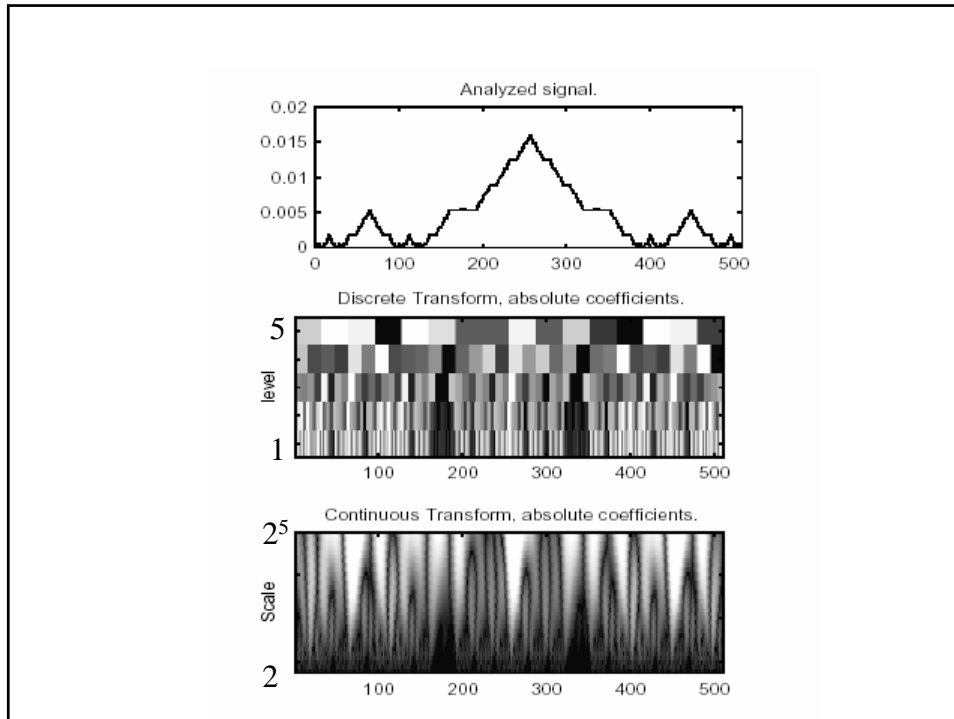
$$\psi_{j,k}(t) = \frac{1}{\sqrt{2^j}} \psi\left(\frac{t - k 2^j}{2^j}\right) = 2^{-j/2} \psi(2^{-j} t - k)$$

$$C(s, d) = C(j, k) = \sum_{n=-\infty}^{\infty} f(n) 2^{-j/2} \psi(2^{-j} n - k)$$

Levels and Resolution

- Scale s and level j are related by: $s = 2^j$
- Resolution: $1/s$
- The smaller is the resolution (larger scale) the higher is the level of detail than can be accessed.

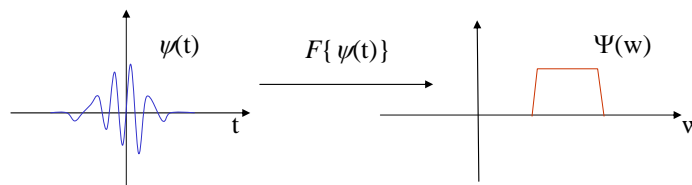
j	10	9	...	2	1	0	-1	-2
Scale	1024	512	...	4	2	1	1/2	1/4
Resolution	$1/2^{10}$	$1/2^9$...	1/4	1/2	1	2	4



Fourier Transform of the Wavelet signal

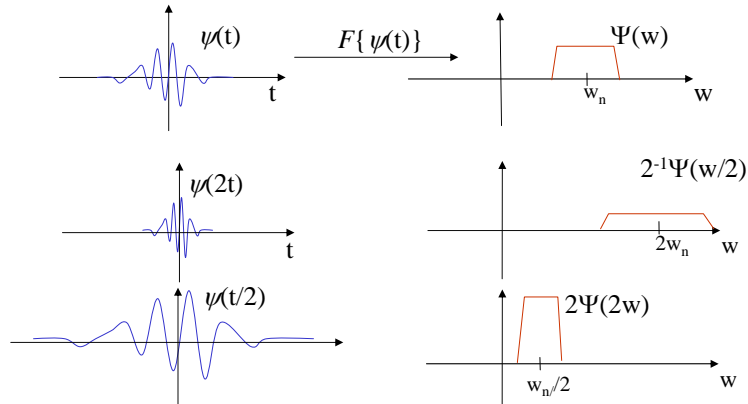
Wavelets have a bandpass structure:

$$\int \frac{|\Psi(w)|^2}{|w|} dw < \infty, \Rightarrow |\Psi(0)|^2 = 0$$



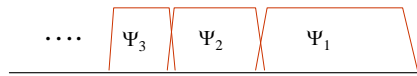
Effect of the Scaling in the FT

$$F\{\psi(at)\} = \frac{1}{|a|} \Psi\left(\frac{w}{a}\right)$$



The wavelet compressed by a factor of 2 extends the frequency of the wavelet spectrum by the same factor and moves the frequencies that factor

- Given a signal we can cover its full spectrum with the spectrum of scaled wavelets in the same way that we can cover the signal in the time domain with the displacement of wavelets.
- If we see a wavelet as a bandpass filter, a series of scaled wavelets can be seen as a bank of bandpass filters.



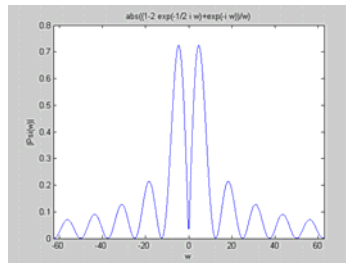
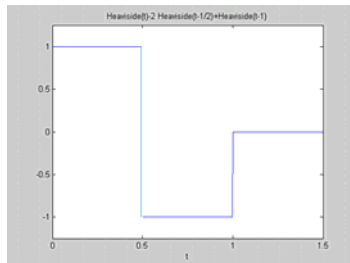
- We can see the wavelet transform of a signal as the signal passing through a bank of filters: Filter Banks

Wavelet de Haar

$$\psi(t) = \begin{cases} 1 & \text{si } 0 \leq t < 1/2 \\ -1 & \text{si } 1/2 \leq t < 1 \\ 0 & \text{en otro caso} \end{cases}$$

Fourier Transform

$$\Psi(w) = \frac{j}{\sqrt{2}} \frac{\left(1 - 2e^{-j\frac{w}{2}} + e^{-jw}\right)}{w}$$

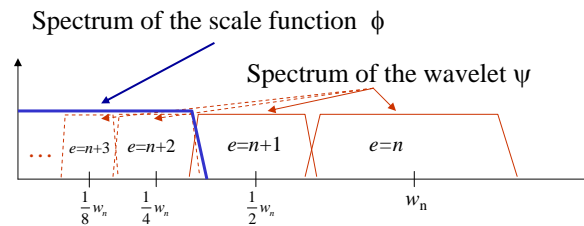


```
% Haar Wavelet
syms w t
wv=sym('Heaviside(t)')-2*sym('Heaviside(t-1/2)')...
      +sym('Heaviside(t-1)');
figure(1),ezplot(wv,[0,1.5])

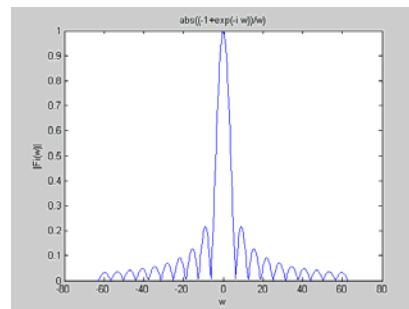
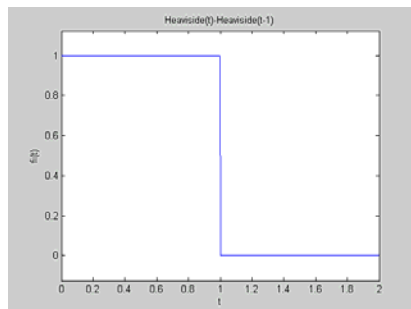
% Its Fourier Transform
WV=fourier(wv); WV=simplify(WV)
figure(2),ezplot(abs(WV),[-20*pi,20*pi])
```

Scale Function

- Everytime we stretch the wavelet by a factor of 2, the bandwidth is halved: reduces by one half: an infinite number of Spectra is needed to reach the zero frequency.
- We make no attempt to cover the entire spectrum, but we use a low pass filter covering the hole, when this is sufficiently small: the spectrum correspond to the denominated *scale function*.



Scale Function for the Haar Wavelet



```
%Haar Scale Function
syms w t
wv=sym('Heaviside(t)')-sym('Heaviside(t-1)');
figure(1),ezplot(wv,[0,2])
% Transformada de Fourier
WV=fourier(wv); WV=simplify(WV)
figure(2),ezplot(abs(WV),[-20*pi,20*pi])
```

Solution of the Refinement Equation

$$\phi(t) = 2 \sum_{k=0}^N h(k) \phi(2t - k)$$

$$\begin{aligned} \int_{-\infty}^{\infty} \phi(t) e^{-j\omega t} dt &= 2 \sum_{k=0}^N h(k) \int_{-\infty}^{\infty} \phi(2t - k) e^{-j\omega t} dt \\ &= 2 \sum_{k=0}^N h(k) \frac{1}{2} \int_{-\infty}^{\infty} \phi(\tau) e^{-j\omega(\tau+k)/2} d\tau \\ &= \sum_{k=0}^N h(k) e^{-j\omega k/2} \int_{-\infty}^{\infty} \phi(\tau) e^{-j\omega\tau/2} d\tau \end{aligned}$$

then

$$\begin{aligned} \Phi(\omega) &= H\left(\frac{\omega}{2}\right) \Phi\left(\frac{\omega}{2}\right) = H\left(\frac{\omega}{2}\right) H\left(\frac{\omega}{4}\right) \Phi\left(\frac{\omega}{4}\right) \\ &\quad \vdots \\ &= \left(\prod_{j=1}^{\infty} H\left(\frac{\omega}{2^j}\right) \right) \Phi(0) \end{aligned}$$

If the scale function area is normalized to one :

$$\Phi(0) = \int_{-\infty}^{\infty} \phi(t) dt = 1$$

$$\Phi(\omega) = \prod_{j=1}^{\infty} H\left(\frac{\omega}{2^j}\right)$$

Relationship between
Filter and Scale Function

Interesting Properties for $H(w)$

- $H(0) = 1$, so $\Phi(0) = 1$
- In order the scale function have finite energy

$$\int_{-\infty}^{\infty} |\Phi(w)|^2 dw < \infty$$

it must be fulfilled: $H(w) \xrightarrow{w \rightarrow \pi} 0$

Solution of the Wavelet Equation

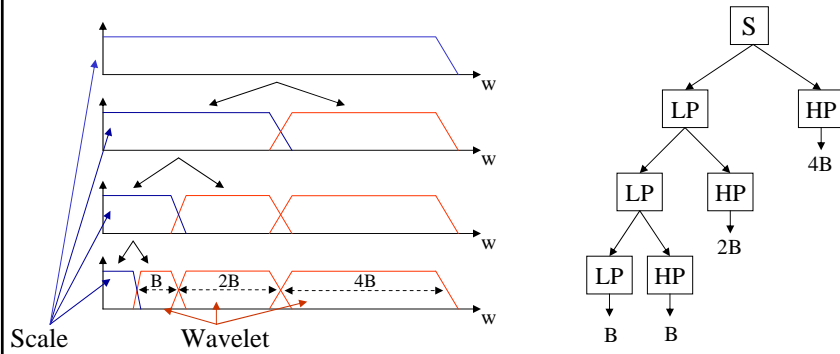
$$\psi(t) = 2 \sum_{k=0}^N g(k) \phi(2t - k)$$

proceeding in a similar way:

$$\begin{aligned} \Psi(w) &= G\left(\frac{w}{2}\right) \Phi\left(\frac{w}{2}\right) \\ &= G\left(\frac{w}{2}\right) \left(\prod_{j=2}^{\infty} H\left(\frac{w}{2^j}\right) \right) \Phi(0) \end{aligned}$$

Band Coding

- We break down the spectrum of the signal in two : a Low Pass (LP) and a High Pass (HP)
 - The HP contains the details, low scale, and the LP contains the approximations, high scale.
- The LP breaks down again in two: LP and HP
- The process continuous until obtaining the desired number of bands

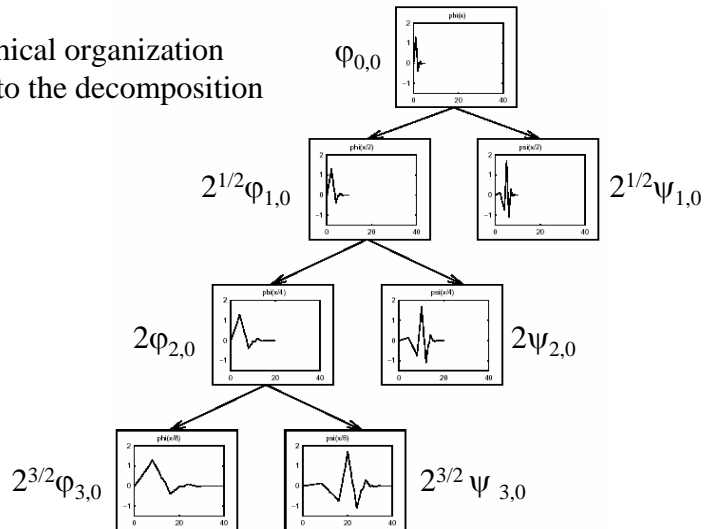


With scales in powers of 2:

$$s = 2^j, \quad d = k 2^j = k s$$

The scale function becomes: $\varphi_{j,k}(t) = \frac{1}{\sqrt{2^j}} \varphi\left(\frac{t - k 2^j}{2^j}\right) = 2^{-j/2} \varphi(2^{-j}t - k)$

Hierarchical organization similar to the decomposition

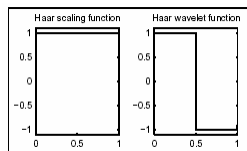


Conclusion

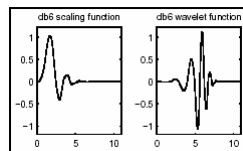
- Applying a Wavelet Transform is equivalent to passing the signal through a bandpass Filter Bank
- The Wavelet defines the details: that is, it gives the bandpass filters with a bandwidth that is reduced by half in each step.
- The scale function defines the approximations.
- If the transformation is performed in this way is not necessary to specify the wavelet explicitly.

Example of Wavelets and their Associated Scale Functions

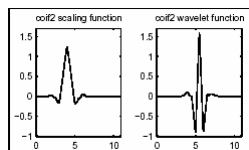
Haar



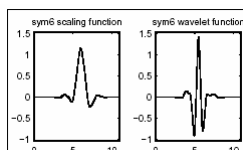
Daubechies

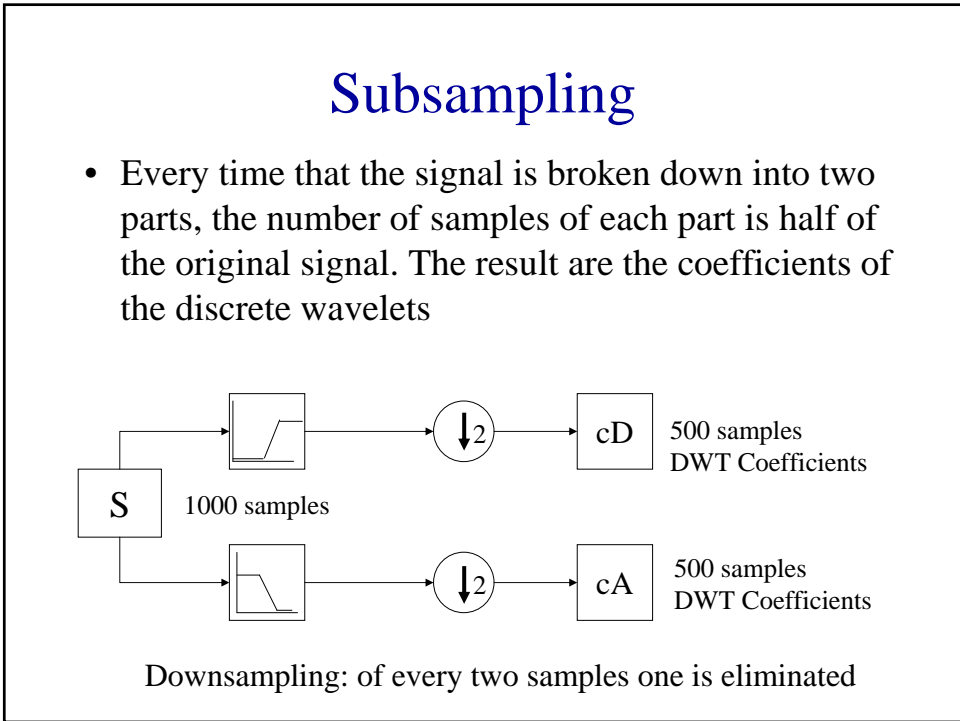
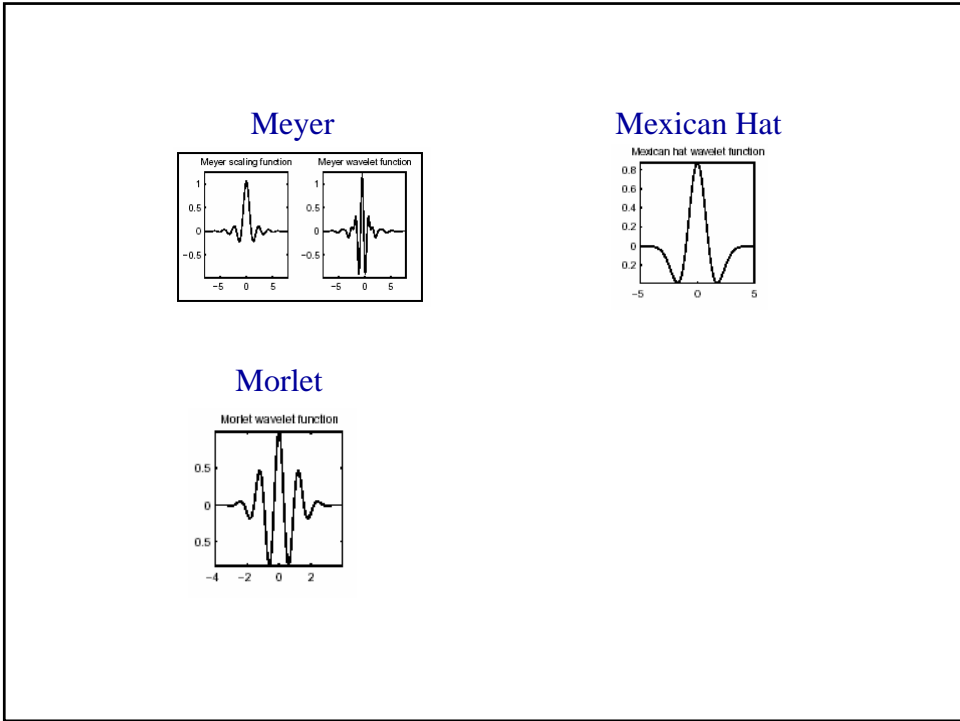


Coiflets



Symlets

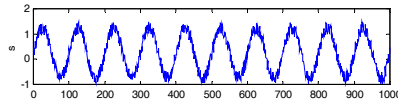




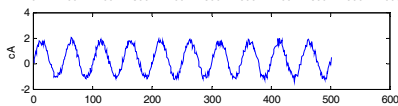
Example 1

```
t=linspace(0,pi,1000);  
s=sin(20*t)+0.5*rand(1,1000);  
[cA,cD]=dwt(s,'db2');  
subplot(3,1,1),plot(s),ylabel('s')  
subplot(3,1,2),plot(cA),ylabel('cA')  
subplot(3,1,3),plot(cD),ylabel('cD')
```

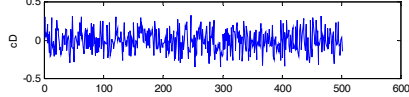
Signal



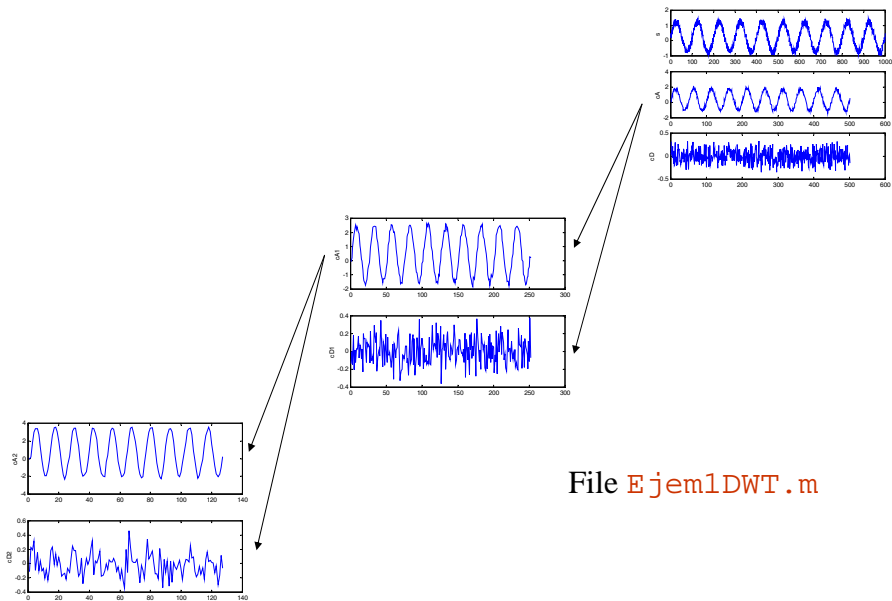
Aproximation
components



Detail
components



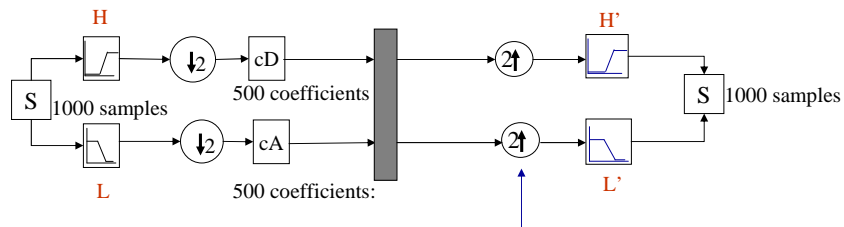
Decomposition Tree



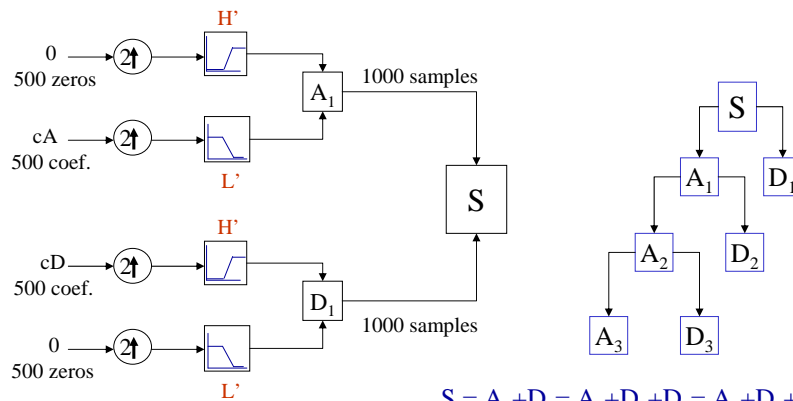
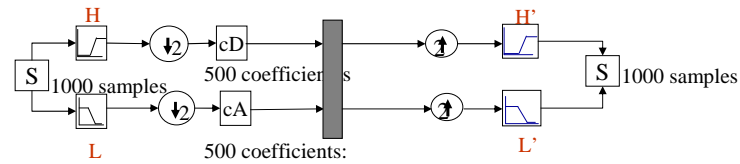
Reconstruction or Synthesis

- Inverse Discrete Wavelet Transform (IDWT)

$$s(t) = \sum_{j=-\infty}^{\infty} \sum_{k=-\infty}^{\infty} C(j,k) \psi_{j,k}(t)$$



Upsampling: lengthening a signal component by inserting zeros between samples.



$$S = A_1 + D_1 = A_2 + D_2 + D_1 = A_3 + D_3 + D_2 + D_1$$

2-Dimensional Transform

For image processing: $\psi(x,y)$

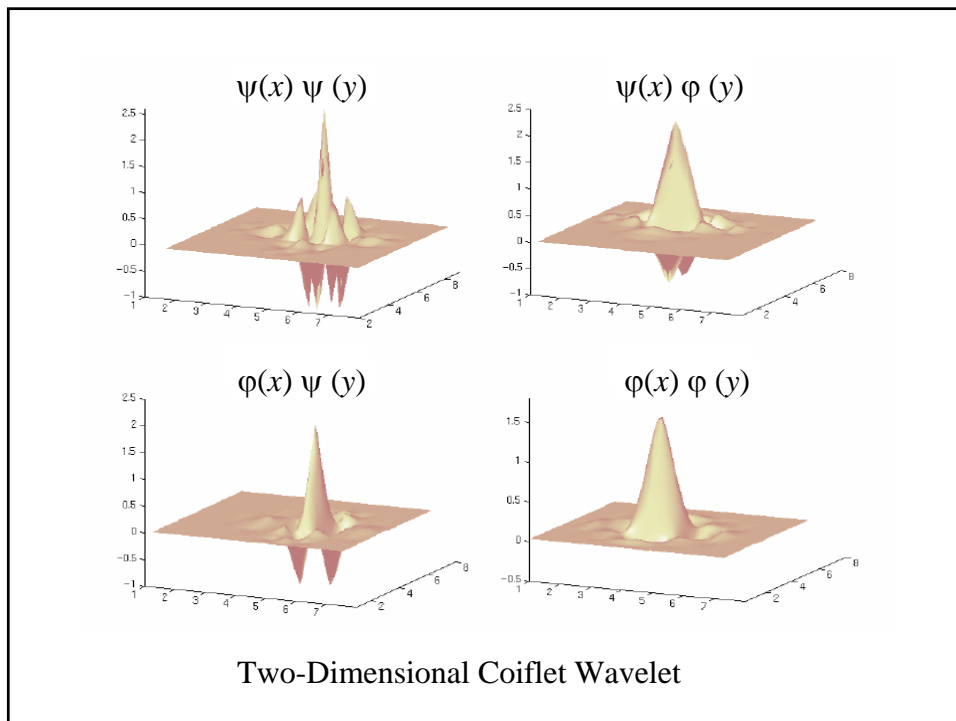
Shifting and Scaling

$$\psi_{s_1, s_2, d_1, d_2}(x, y) = \frac{1}{\sqrt{s_1 s_2}} \psi\left(\frac{x-d_1}{s_1}, \frac{y-d_2}{s_2}\right)$$

Scale values in powers of 2: $s = 2^j$, $d = k 2^j = k s$

Two-dimensional wavelet is defined as the tensor product of 1-dimensional wavelets:

Scale Function:	$\varphi(x,y) = \varphi(x) \varphi(y)$
Wavelets:	$\psi_1(x,y) = \varphi(x) \psi(y),$
	$\psi_2(x,y) = \psi(x) \varphi(y),$
	$\psi_3(x,y) = \psi(x) \psi(y)$



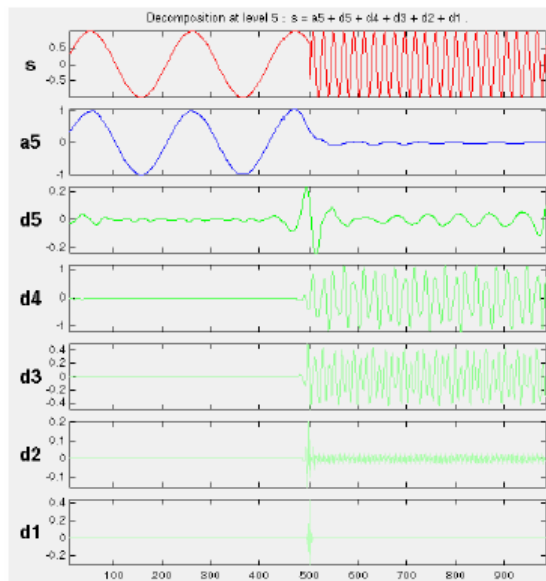
Applications

- Detecting Discontinuities
- Detecting Trends
- Detecting Self-Similarity
- Identifying Pure Frequencies
- Suppressing Signals
- De-Noising Signals
- Compressing Signals

Detecting Discontinuities

Dos sinusoides de distintas frecuencias

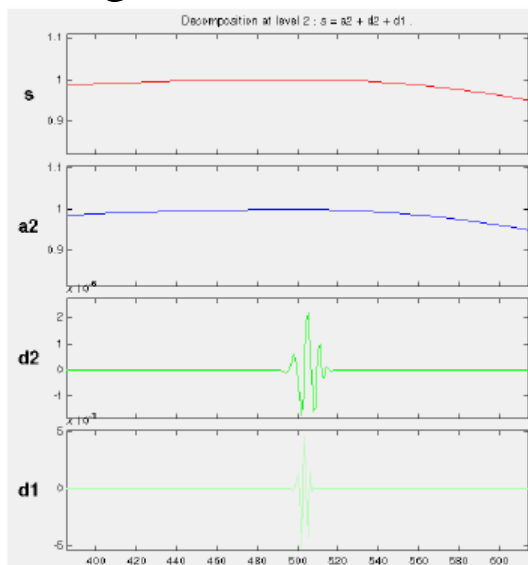
Wavelet: db5
Nivel: 5



Detecting Discontinuities

Dos exponenciales
conectadas
en $t = 500$

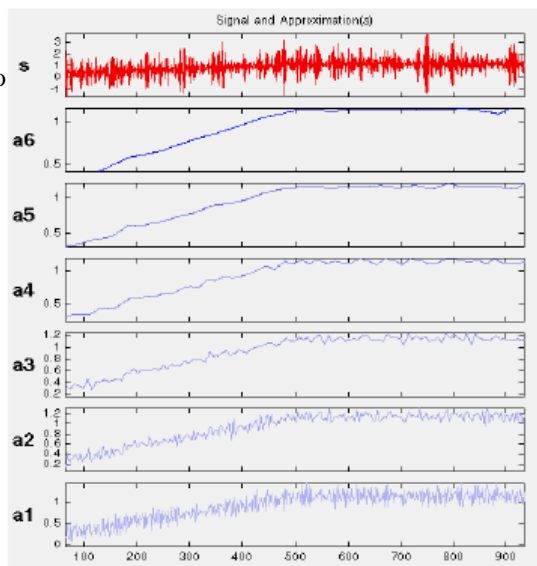
Wavelet: db4
Nivel: 2



Detecting Trends

Rampa oscurecida
por ruido coloreado

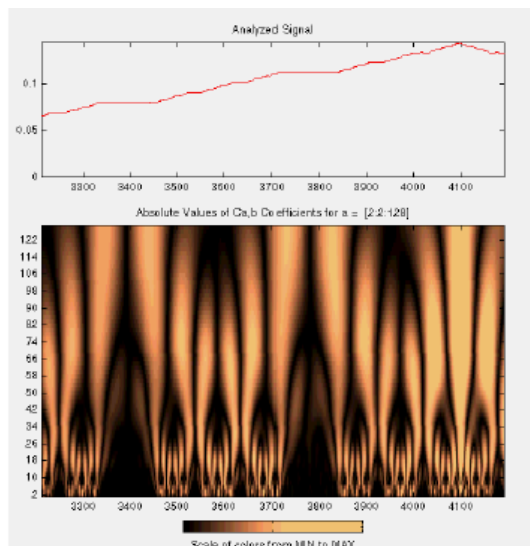
Wavelet: db3
Nivel: 6



Detecting Self-Similarity

Fractal

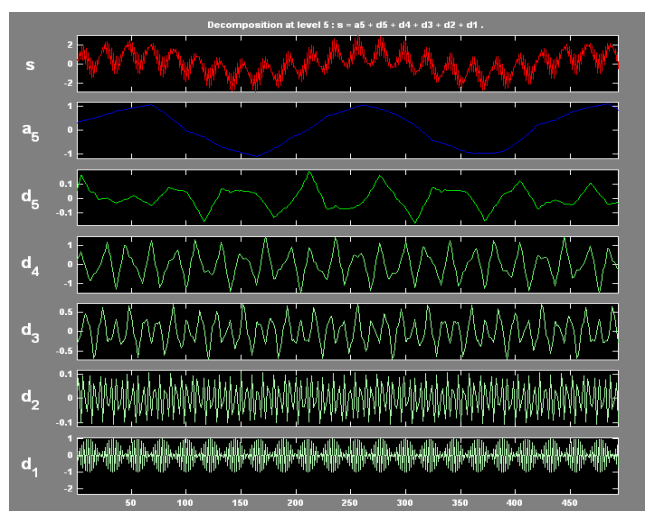
Wavelet: coif3
Nivel: 2:2:128



Detecting Frequencies

Tres sinusoides de
Frecuencias:
0.005, 0.05, 0.5
(A₅, D₄, D₁)

Wavelet: db3
Nivel: 5



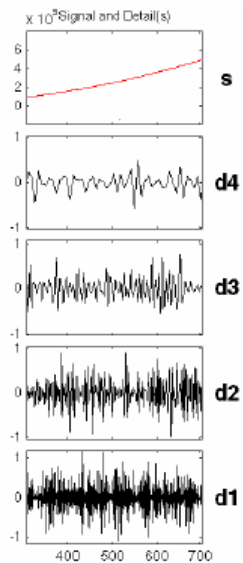
Suppressing Signals

Polinomio de 2º orden con ruido.

En los detalles se ha eliminado completamente la señal polinómica

Wavelet: db3

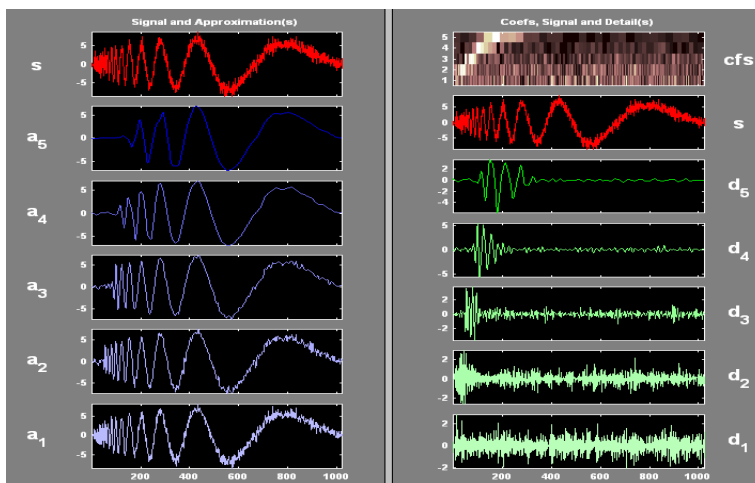
Nivel: 4



De Noising Signals

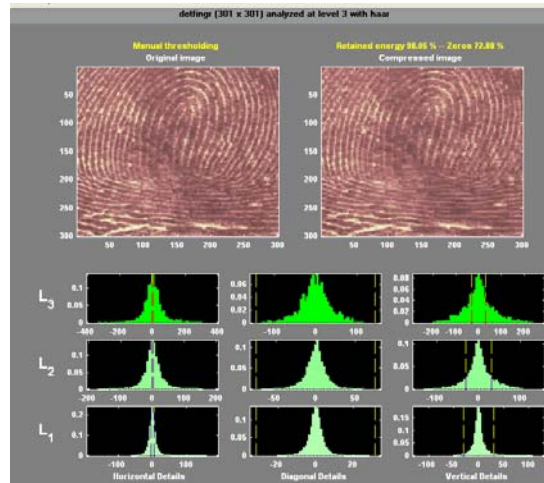
Desplazamiento Doppler de una senoide con adición de ruido

Wavelet: sym4, Nivel: 5



Compressing Images

Usada en la compresión de imágenes para almacenamiento de información.



Summary

The wavelets allow you to perform stationary analysis of signals, as the Fourier transform does, but also analysis of localized areas of a signal, allowing study characteristics such as shifts, trends, abrupt changes and start and end events.

Wavelet Transform can be interpreted as a Filter Banks, so its explicit specification is not required.

Two-dimensional Wavelet Transform can be used for image processing.

Bibliography

Bogges, A., F.J. Narcowich (2001), "A First Course in Wavelets with Fourier Analysis", Prentice Hall.

Daubechies, I. (1992), "Ten lectures on wavelets", SIAM.

Strang, G., T. Nguyen (1996), "Wavelets and filter banks", Wellesley-Cambridge Press.

Mallat, S. (2001), "A wavelet tour of signal processing". 2^o Edition, Academic Press.

<http://www.wavelet.org/>

Wavelet Toolbox for use with MATLAB.